

Mobile Communications (KECE425)

Lecture Note 5

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Summary

- Outage probability
 - Area outage probability
- Multi-path fading

Outage Probability

- Carrier-to-noise ratio

$$\Gamma = \frac{\text{Carrier power}}{\text{Noise power}}$$

- Thermal noise outage probability

$$O_N = \Pr[\Gamma < \Gamma_{\text{th}}]$$

- Carrier-to-interference ratio

$$\Lambda = \frac{\text{Carrier power}}{\text{Interference power}}$$

- Co-channel interference outage probability

$$O_I = \Pr[\Lambda < \Lambda_{\text{th}}]$$

- Overall outage due to both thermal noise and co-channel interference

$$O = \Pr[\Gamma < \Gamma_{\text{th}} \text{ or } \Lambda < \Lambda_{\text{th}}]$$

- Edge noise outage probability

$$\begin{aligned} O_N(R) &= P(\Omega_{p(\text{dBm})}(R) < \Omega_{\text{th}(\text{dBm})}) \\ &= \int_{-\infty}^{\Omega_{\text{th}(\text{dBm})}} \frac{1}{\sqrt{2\pi}\sigma_{\Omega}} \exp\left\{-\frac{(x - \mu_{\Omega_{p(\text{dBm})}}(R))^2}{2\sigma_{\Omega}^2}\right\} dx \\ &= Q\left(\frac{M_{\text{shad}}}{\sigma_{\Omega}}\right) \end{aligned}$$

where $M_{\text{shad}} = \mu_{\Omega_{p(\text{dBm})}} - \Omega_{\text{th}(\text{dBm})}$ is called *Shadow margin*.

- Example

- Suppose that we wish to have $O_N(R) = 0.1$. Determine the Shadow margin M_{shad} .

- Solution

- We solve

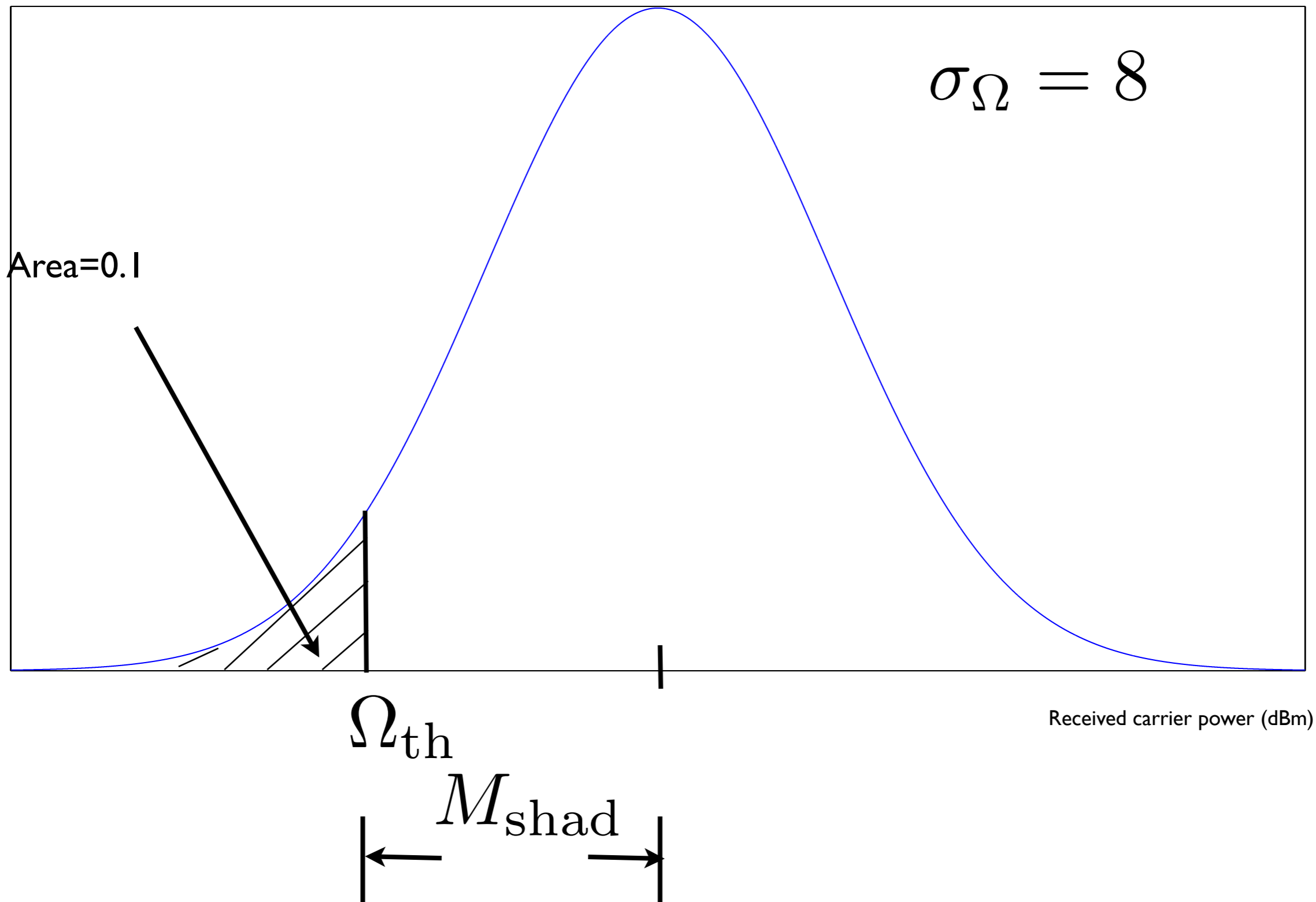
$$0.1 = Q\left(\frac{M_{\text{shad}}}{\sigma_{\Omega}}\right),$$

which gives

$$\frac{M_{\text{shad}}}{\sigma_{\Omega}} = Q^{-1}(0.1) = 1.28$$

- For $\sigma_{\Omega} = 8$ dB, the required shadow margin is

$$M_{\text{shad}} = 1.28 \times 8 = 10.24 \text{ dB}$$



Area Outage Probability

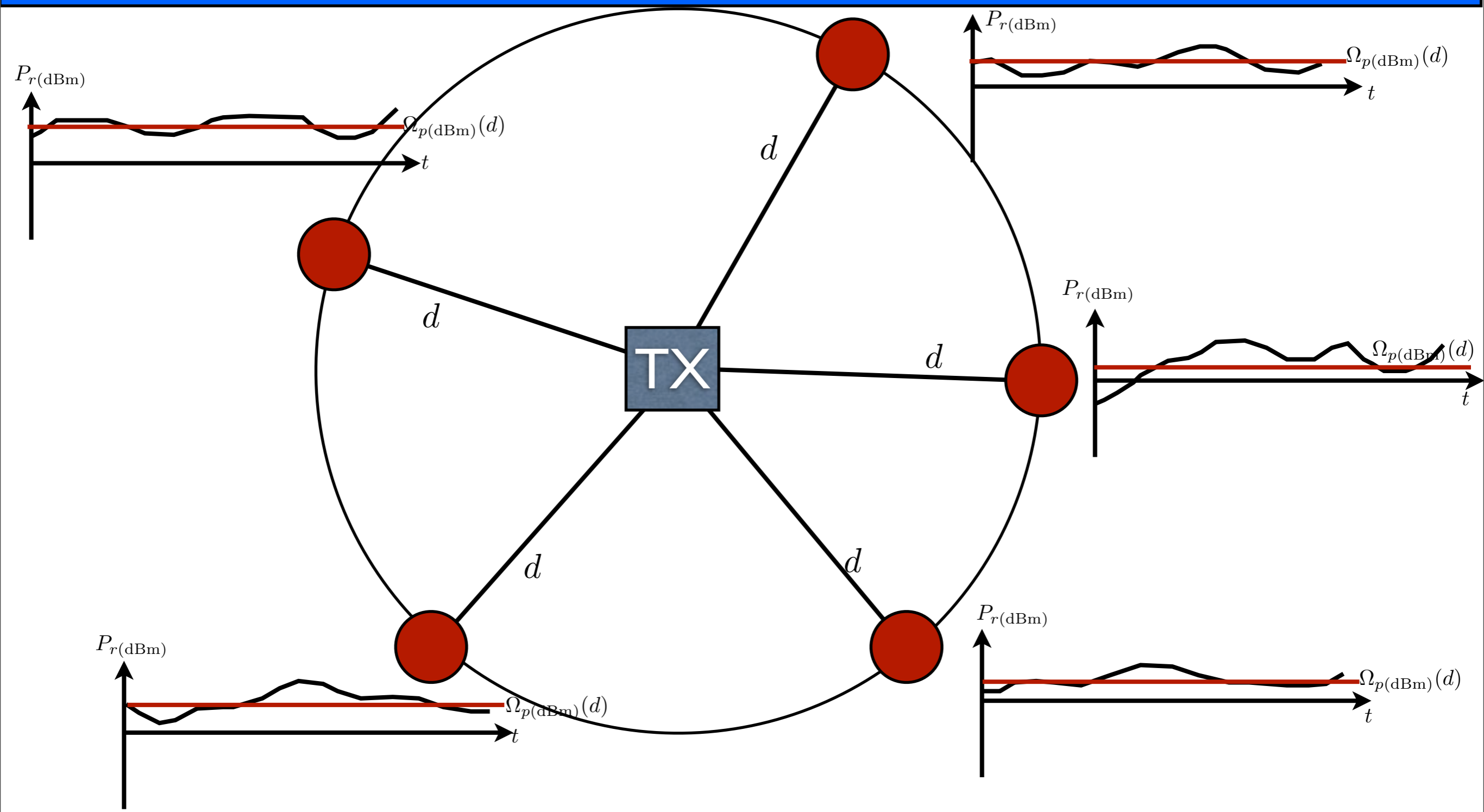
- Area outage probability averaged over area of a cell:

$$\begin{aligned} O_N &= \frac{1}{\pi R^2} \int_0^R O(r) 2\pi r dr \\ &= Q(X) - \exp\{XY + Y^2/2\} Q(X + Y) \end{aligned}$$

where

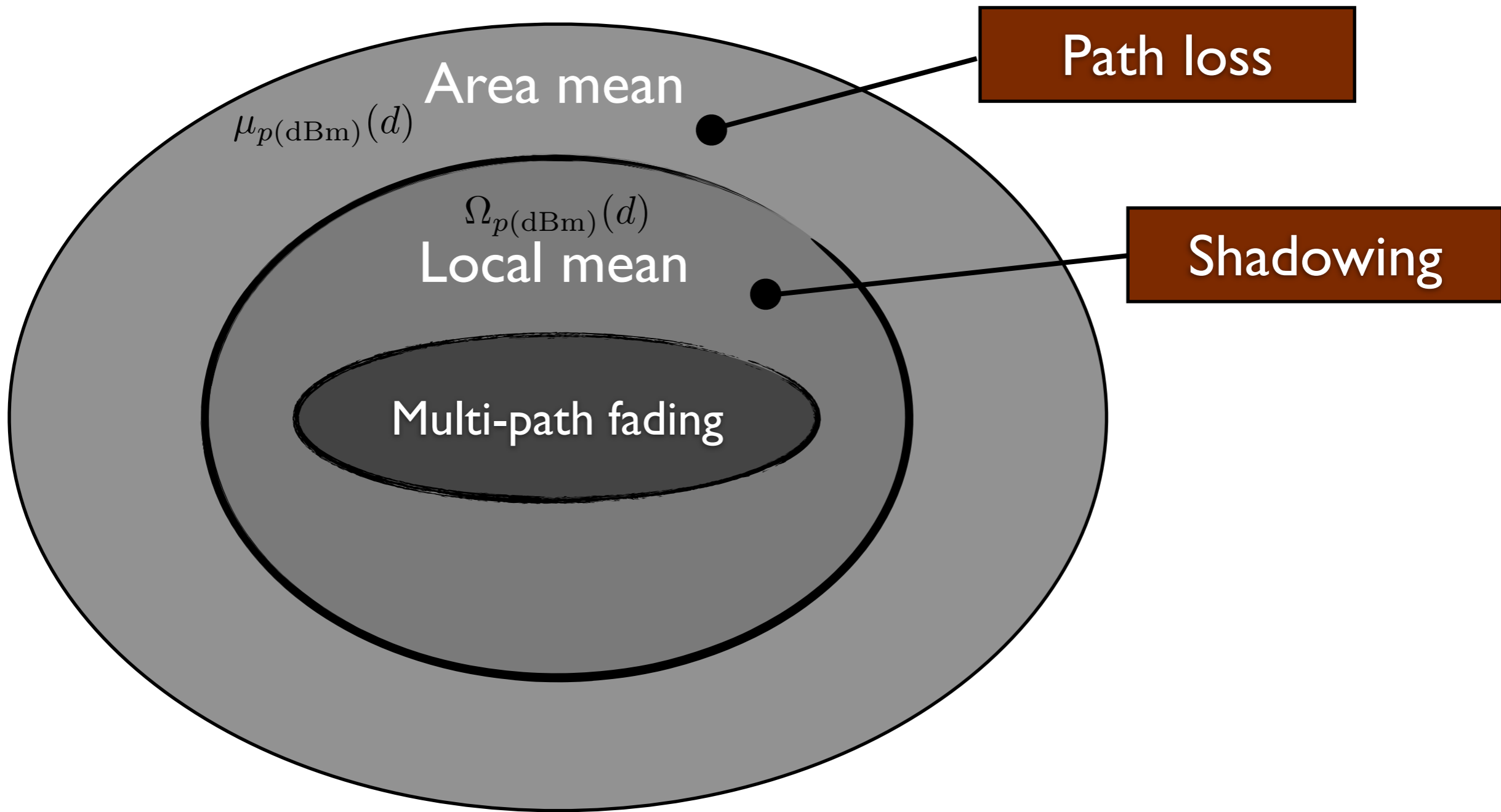
$$\begin{aligned} X &= \frac{M_{\text{shad}}}{\sigma_\Omega} \\ Y &= \frac{2\sigma_\Omega}{\beta\zeta} \\ \zeta &= \frac{10}{\ln 10} \end{aligned}$$

Wireless Channel Model



Wireless Channel Models

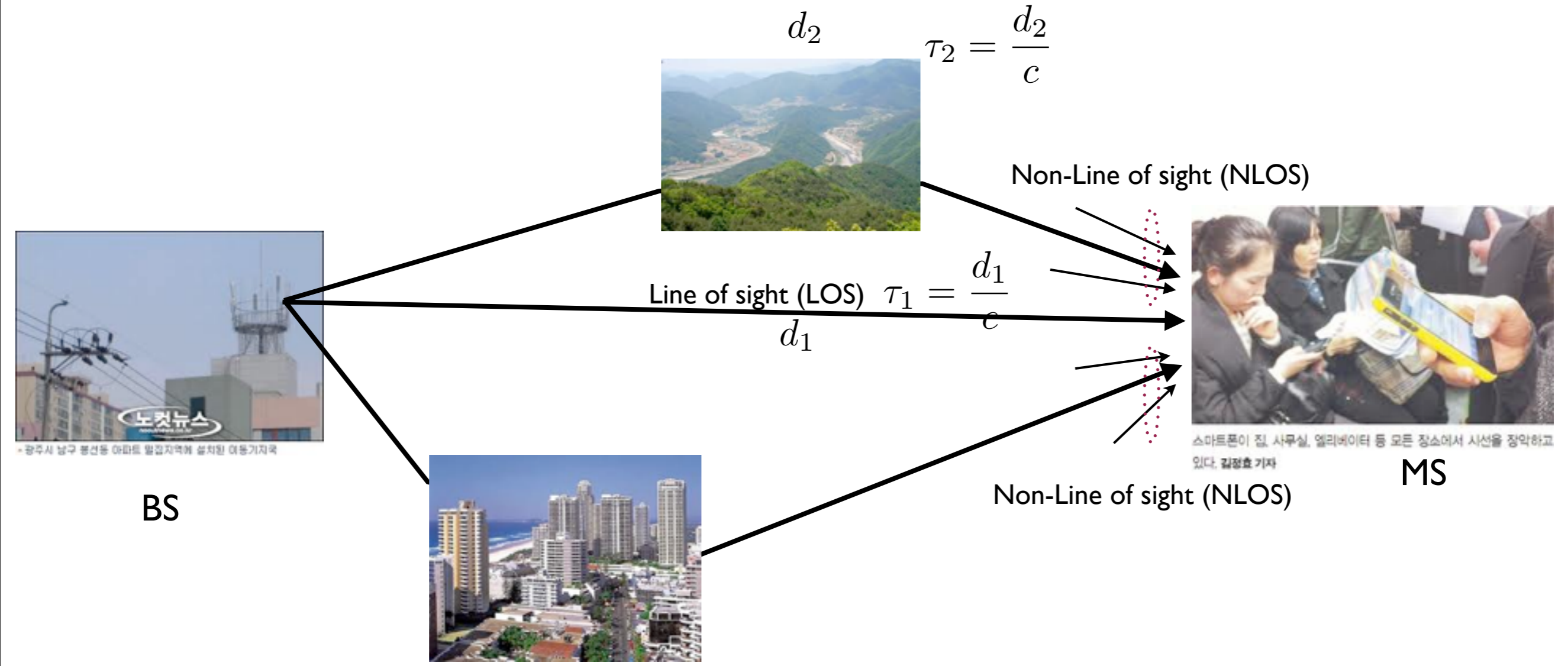
- Large scale channel model
 - Path loss model
 - Shadowing
- Small scale channel model
 - Multi-path fading



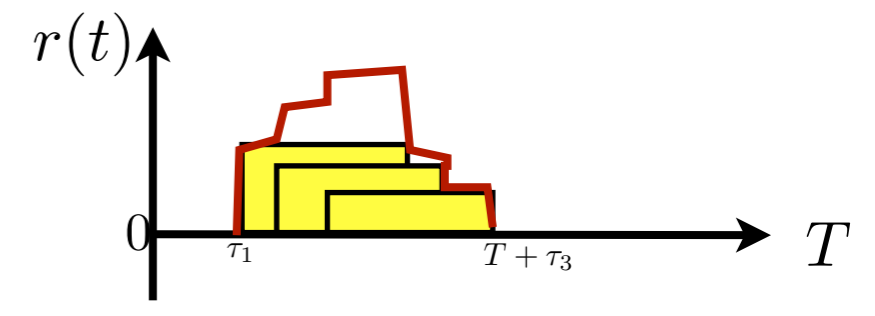
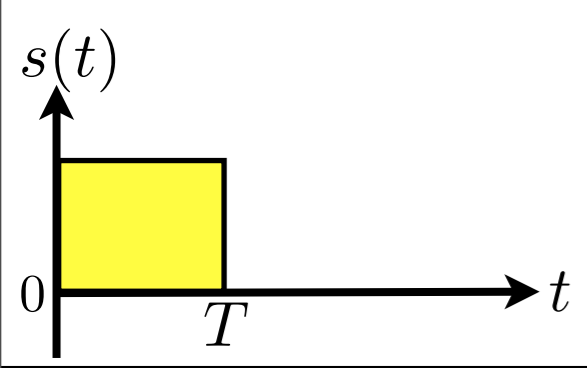
Multi-path Fading

- Categorization of multi-path fading
 - Depending on the vehicle speed
 - Fast fading vs. slow fading
 - Depending on the signal bandwidth (or data rate)
 - Frequency flat fading vs. frequency selective fading

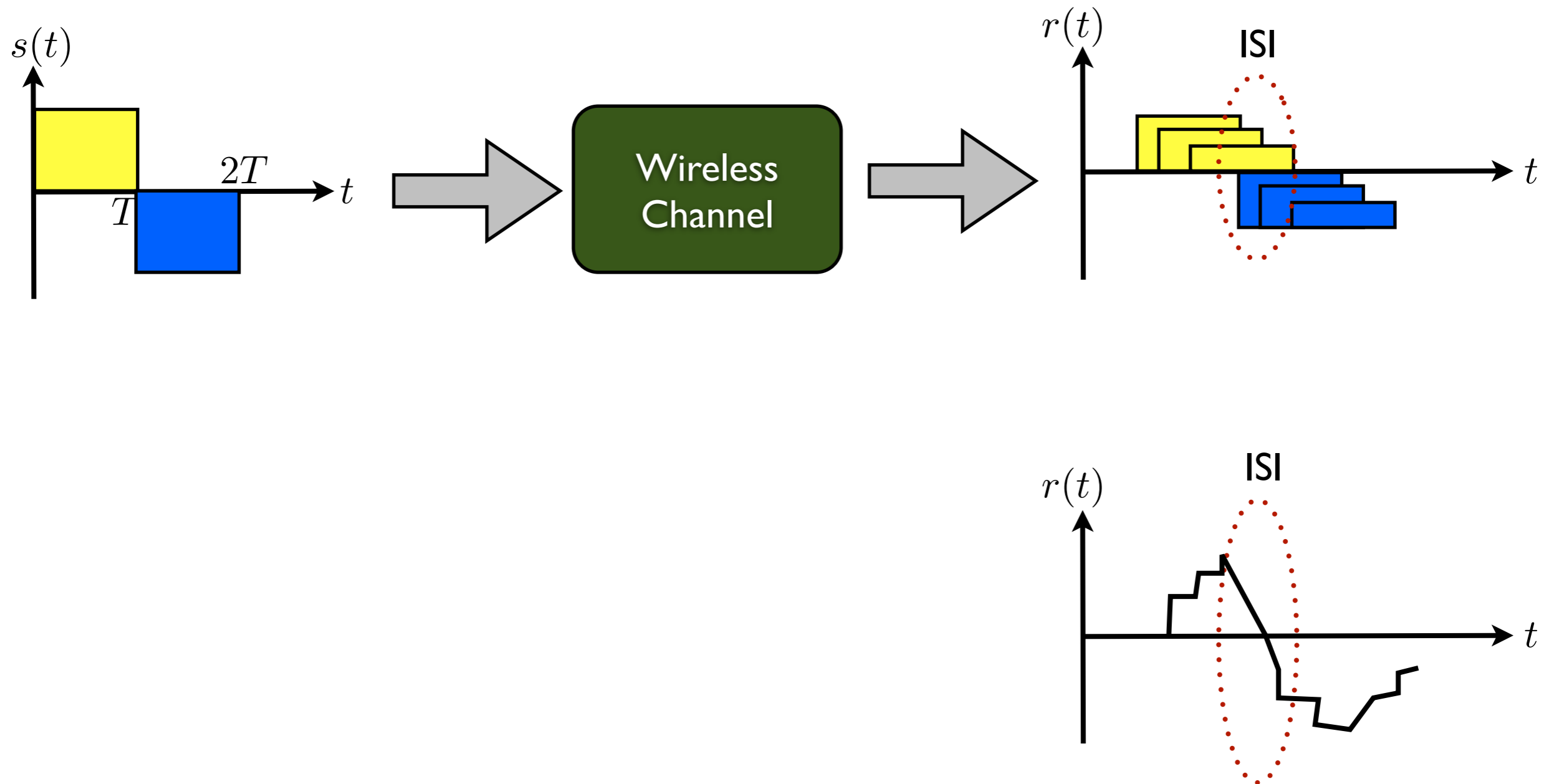
Multi-Path Phenomenon



$$d_1 < d_2 < d_3 \implies \tau_1 < \tau_2 < \tau_3$$

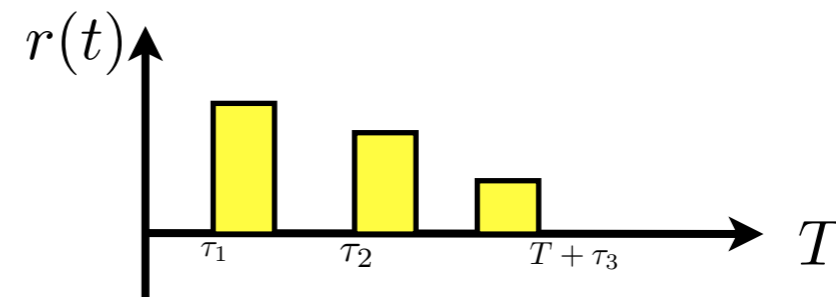
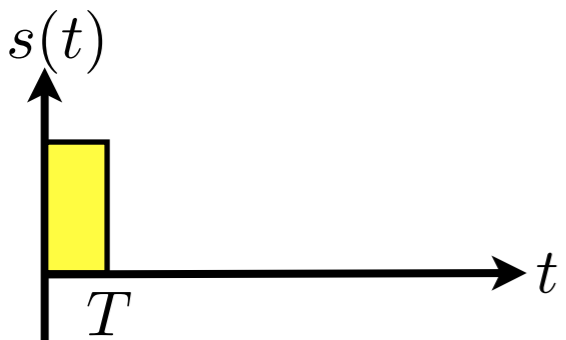
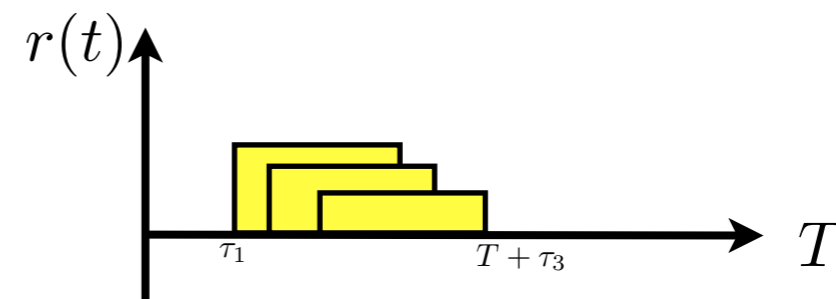
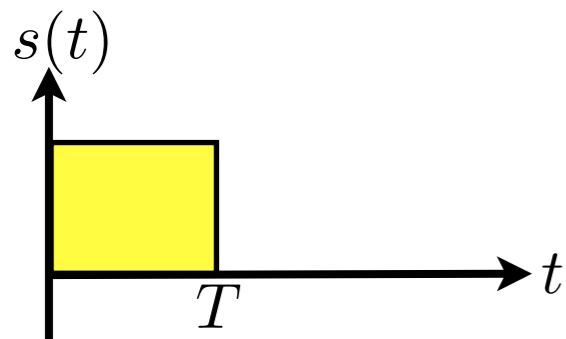


Inter-Symbol Interference due to Multi-Path Fading



Effect of Data Rate (or Bandwidth)

- Smaller time duration of the transmitted signal => higher data rate



ISI is getting severer when the data rate gets higher.

Wireless Channels

Wireless
Channel

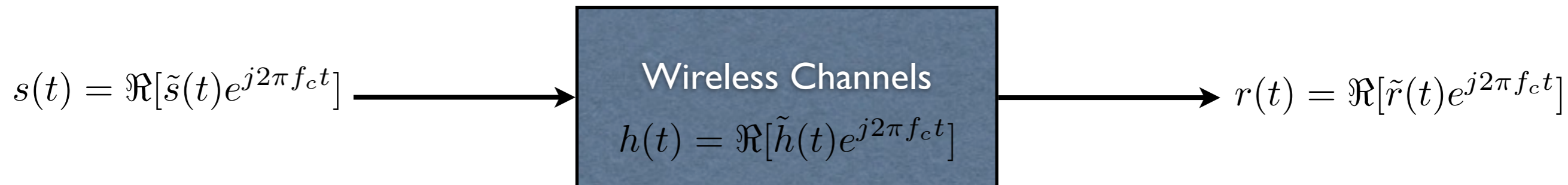


Received signal power

- *Time-varying random signal*
- *multi-path signal*

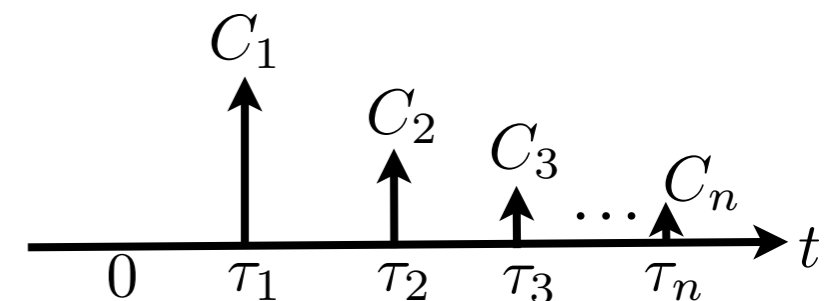
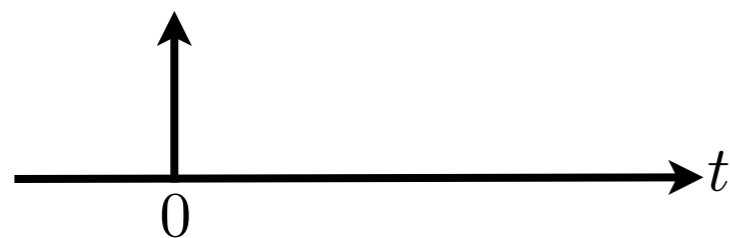
Impulse Response of the Wireless Channels

Equivalent low-pass signal and system representation



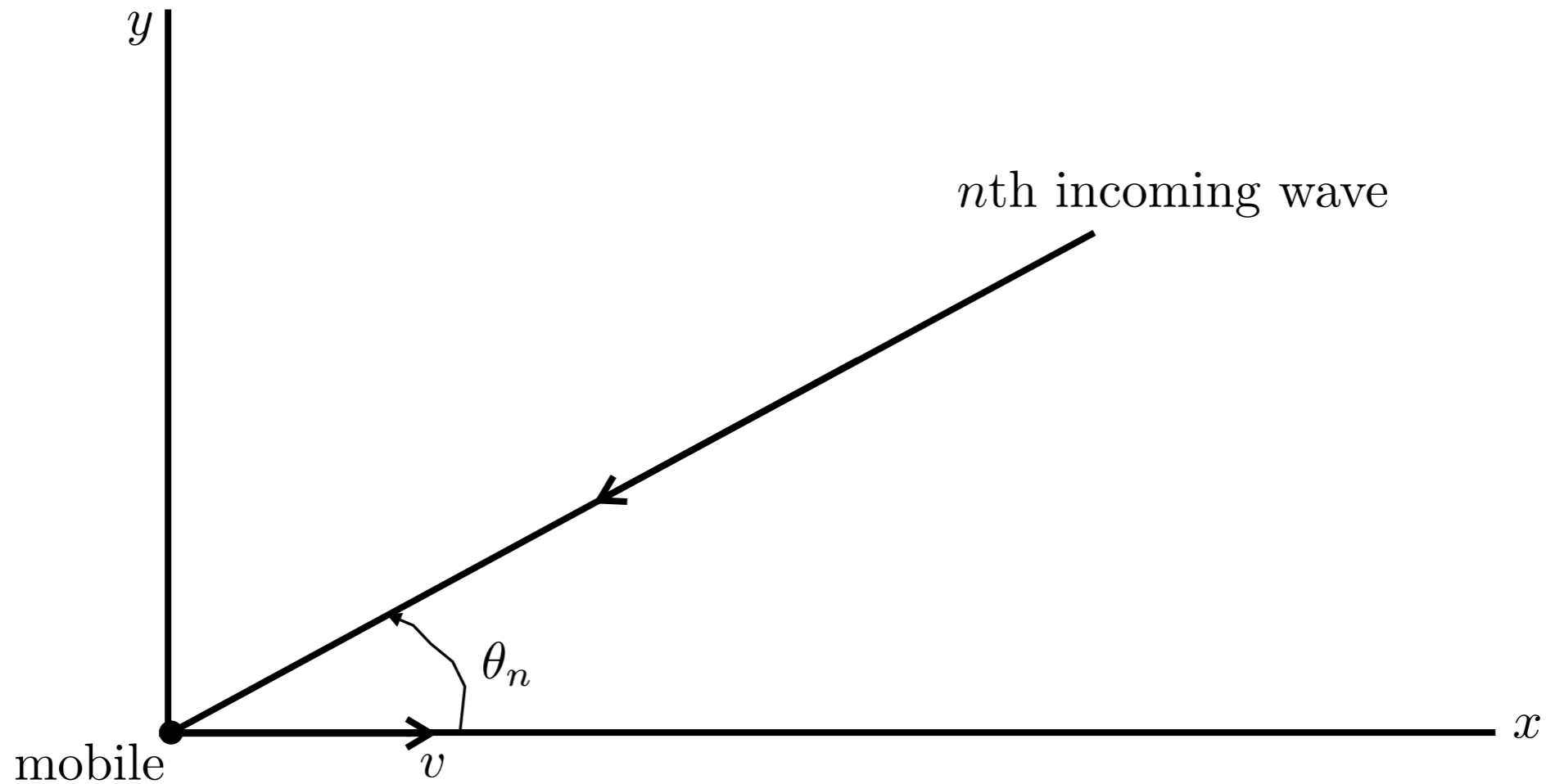
$$r(t) = s(t) * h(t) \quad \tilde{r}(t) = \tilde{s}(t) * \tilde{h}(t)$$

- only if the channel can be modeled as linear-time-invariant system
- However, the wireless channel is time-variant.



In practice, there are many multiple paths arriving at the receiver with random amplitude and phase.

Doppler Effect



Doppler frequency

$$f_{D,n} = f_m \cos \theta_n \text{ Hz}$$

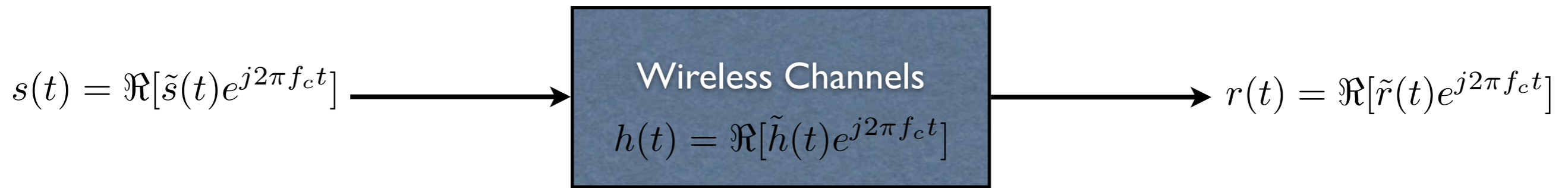
where $f_m = \frac{v}{\lambda_c}$ is maximum Doppler frequency.

Frequency Shift Due to Doppler Effect

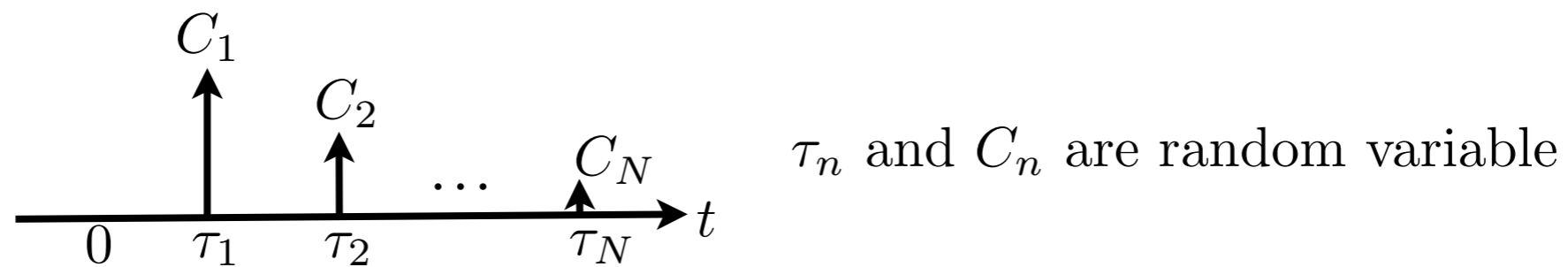
Example: $f_c = 2 \text{ GHz}$ $\lambda_c = 15 \text{ cm}$

Vehicle speed (km/hr)	Maximum Doppler freq. (Hz)	Frequency shift
3	5.56	2GHz + 5.56 Hz
30	55.56	2GHz + 55.56 Hz
60	111.11	2GHz + 111.11 Hz
120	222.22	2GHz + 222.22 Hz
300	555.56	2GHz + 555.56 Hz

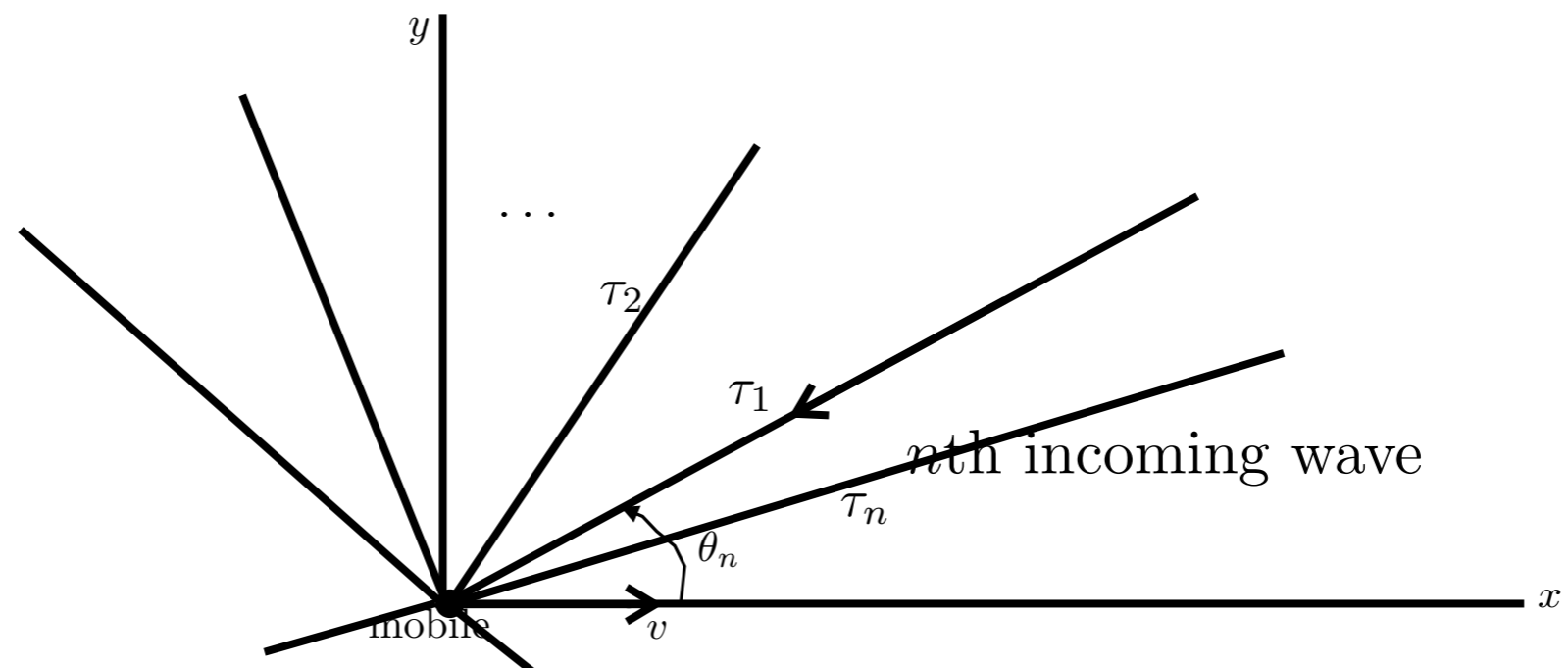
- Doppler frequency shift: $f_c + f_{D,n} = f_c + f_m \cos \theta_n$



$$h(t, \tau)$$



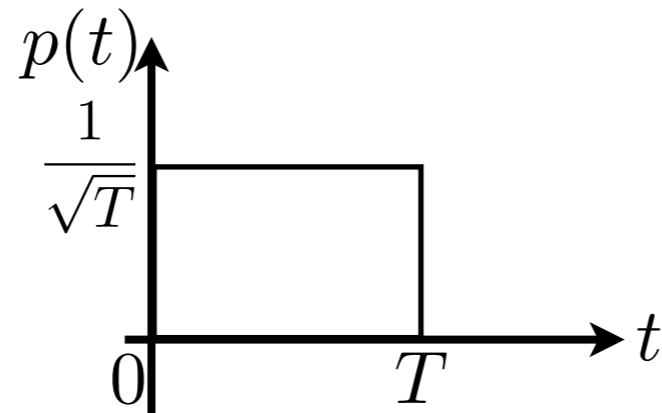
- C_n and τ_n are the amplitude and time delay, respectively associated with the n th propagation path.



- Transmitted signal:

$$s(t) = a_m p(t) \cos(2\pi f_c t + \theta_m), \quad m = 1, 2, \dots, M$$

where $p(t)$ is the pulse shape for the symbol duration.



- Equivalent lowpass signal form:

$$s(t) = \Re [\tilde{s}(t) e^{j2\pi f_c t}]$$

where $\tilde{s}(t) = a_m p(t) e^{j\theta_m}$

- Received signal is the sum of a total of N multi-path signals:
 - First incoming wave: $C_1 a_m p(t - \tau_1) \cos(2\pi(f_c + f_{D,1})(t - \tau_1) + \theta_m)$
 - Second incoming wave: $C_2 a_m p(t - \tau_2) \cos(2\pi(f_c + f_{D,2})(t - \tau_2) + \theta_m)$
 - ⋮
 - n th incoming wave: $C_n a_m p(t - \tau_n) \cos(2\pi(f_c + f_{D,n})(t - \tau_n) + \theta_m)$
 - ⋮
 - N th incoming wave: $C_N a_m p(t - \tau_N) \cos(2\pi(f_c + f_{D,N})(t - \tau_N) + \theta_m)$

